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DESIGN OF AREA- EFFICIENT PRBG ARCHITECTURE USING SUQARE ROOT CARRY SELECT ADDER

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ABSTRACT: Security and privacy over the internet is the most sensitive and primary objective to protect data. The data can be protected by using various cryptography and pseudorandom bit generator (PRBG) techniques. Three-operand binary adder is the basic functional unit to perform the modular arithmetic in various cryptography and pseudorandom bit generator (PRBG) algorithms. Square root carry select adder is used for three-operand addition that significantly reduces the critical path delay and area of the architecture. The three-operand binary addition is thus performed using RCA logics and multiplexers, which is a novel area-efficient adder architecture that uses far less space by decreasing the number of Look Up Tables, less power, and significantly less adder delay. This proposed adder is placed in the every LCG block of MDCLG Architecture. Here, MDCLG generates a 32-bit random code. Moreover, it has a lesser area and lower power dissipation. Also, the proposed adder achieves less area than the existing three-operand adder techniques.

KEYWORDS: Cryptography, PRBG, Square Root Carry Select Adder, RCA, LCG, MDCLG.

I.INTRODUCTION

To enhance the development in VLSI, the cryptography methods must be implemented on software in order to obtain the best system performance while maintaining confidentiality. The arithmetic operations in different cryptographic methods typically involve modular arithmetic, such as modular exponentiation, modular multiplication and modular addition. As a result, the effectiveness of the congruential modular arithmetic operation depends on the cryptographic algorithm which was implemented. The Montgomery algorithm, whose key operation is based on three-operand binary addition, is the most effective method for implementing modular multiplication and exponentiation. In Linear Congruential Generator (LCG) based Pseudo-Random Bit Generators (PRBG), such as linked LCG (CLCG), modified dual-CLCG (MDCLCG), and coupled variable input LCG (CVLCG), the threeoperand binary addition is also a key arithmetic operation. Smaller area and less power are becoming key design considerations for VLSI circuits in the constantly expanding electronic industry, in addition to quicker units. Hence in order to increase the mobility and battery life of portable devices, a VLSI designer must optimise area delay and power limits. While millions of instructions per second are executed by microprocessors, operating speed is the most crucial restriction to take into account when constructing multipliers. Due to the difficulty of achieving these limitations, a compromise between them must be established depending on the application. Adder design can be done in a variety of ways.

By balancing the delay through two carry chains and the block multiplexer signal from the previous stage, the squareroot carry select adder is built. Non-linear carry select adder is another name for it. An arithmetic combinational logic circuit known as a carry select adder adds two N-bit binary values and outputs the resulting N-bit binary sum and a 1bit carry. The fundamental benefit of CSA is its characteristics of reduced propagation delay. This is accomplished by the use of parallel stages, which are produced by using many pairs of ripple carry adders. The carry input is assumed to be 0 or 1, respectively, by the ripple carry adders, who provide the intermediate sum and carry for the CSA structure.

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II.EXISTING SYSTEM

In order to accomplish the three-operand addition in modular arithmetic, this section introduces a new adder technique and associated VLSI design. The adder method that is being suggested is a parallel prefix adder. To compute the addition of three binary input operands, such as bit-adding logic, base logic, PG (propagate and generate) logic, and sum logic, it has four-stage structures rather than three-stage structures in the prefix adder. The definitions of each of these four levels' logical expressions are as follows.

 $c_{n-1}b_{n-1}a_{n-1}$

FFF

Stage-1: Bit Addition Logic:

HHH =

$$S'_{i} = a_{i} \oplus b_{i} \oplus c_{i},$$

$$cy_{i} = a_{i} \cdot b_{i} + b_{i} \cdot c_{i} + c_{i} \cdot a_{i}$$

Stage-2: Base Logic

$$G_{i:i} = G_i = S'_i \cdot cy_{i-1}, \quad G_{0:0} = G_0 = S'_0 \cdot C_{in}$$

$$P_{i:i} = P_i = S'_i \oplus cy_{i-1}, \quad P_{0:0} = P_0 = S'_0 \oplus C_{in}$$

Stage-3: PG (Generate and Propagate) Logic

$$S_i = (P_i \oplus G_{i-1:0}), \quad S_0 = P_0, \ C_{out} = G_{n:0}$$

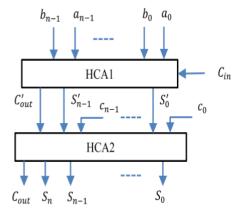
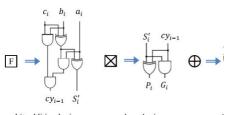
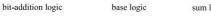
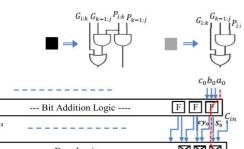


Fig. 2.1: Block level architecture of HCAbased three-operand adder (HC3A).







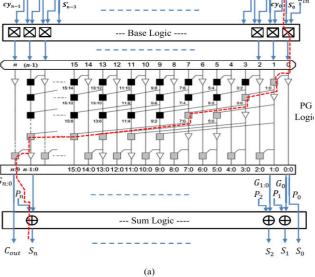


Fig. 2.2: Existed three-operand adder; (a) First order VLSI architecture.

The HCA-based three-operand adder's (HC3A) detailed architecture is shown in. The propagate chain, or the number of blackgrey cell stages in the PG logic of the Han-Carlson adder, determines the maximum combinational path delay of the HC3A and is calculated as follows:

$$T_{\text{HC3A}} \approx 4T_{\text{X}} + 4 \lfloor \log_2 n \rfloor T_{\text{G}}$$
$$A_{\text{HC3A}} \approx (4n+1) A_{\text{X}} + 6 \left[n + \left\lceil \frac{n}{2} \right\rceil s - 2^s + 1 \right] A_{\text{G}}$$

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In comparison to the HCA-based three-operand adder (HC3A) significantly increases the area grows in the order of O with increasing bit length (n log2 n). Thus, to reduce the area in the MDCLG, we chosen the proposed adder such that is Square Root carry select adder.

III.PROPOSED SYSTEM

Two RCA blocks make up the Carry Select Adder. In this project, we suggested a square root carry select adder to shorten the ideal delay. The block size in Square Root Carry Select Adder (SRCSA) might vary. A 16-bit adder can be made utilising block sizes of 2-2-3-4-5 rather than using a uniform block size of four (as done before). The full analysis is omitted here for conciseness. When the Full-Adder delay and the MUX delay are equal, this break-up is perfect. The inputs are A and B, the carry-in is denoted as Cin, and the outputs are denoted by sum (S) and carry-out in the block diagram of the proposed SRCSA adder in Fig.3.1. (Cout).

By balancing the delay through two carry chains and the block multiplexer signal from the previous stage, the squareroot carry select adder is built. Non-linear carry select adder is another name for it.

The primary drawback of conventional CSLA is the enormous space caused by the numerous pairs of ripple carry adders. The basic square-root Carry Select adder features a dual ripple carry adder with a 2:1 multiplexer. Here is a diagram of a standard 16-bit SQRT Carry select adder.

16-bit sqrt carry select adder

A 16-bit sqrt carry-select adder is divided into sectors, each of which, with the exception of the least significant, executes two additions concurrently, one assuming a carry-in of zero and the other a carry-in of one. Additionally, each sector has two 4-bit rcas that receive the same data inputs but different Cin.

One of the two adders is chosen based on the real Data from the sector before. The sum and carryout of the higher adder are chosen if the carry-in is zero. The sum and carry-out of the lower adder are chosen if the carry-in is one. Regular 16-bit SQRT CSLA has a five sets of RCAs of various sizes.

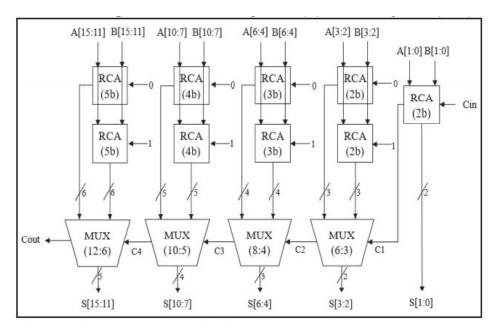


Fig. 3.1: 16-bit SRCSA (proposed adder)



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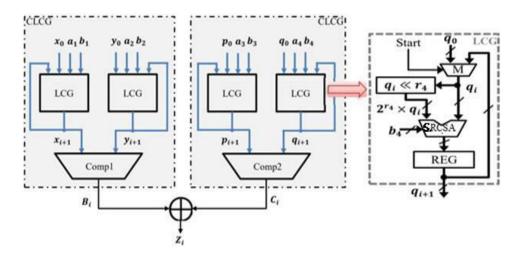
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PERFORMANCE OF THE MODIFIED DUAL-CLCG ARCHITECTURE WITH THE THREE-OPERAND ADDER

For the fastest encryption and decryption, the hardware security in the field of IoT applications requires streamcipher based high data rate, lightweight cryptography technology. The main part of stream-cipher based encryption and decryption is the key generator, also known as the pseudorandom bit generator (PRBG). The most effective PRBG technique that is suited for hardware security based on stream cyphers is modified dual-CLCG (MDCLCG). Yet, the bit size of the congruential modulus has a linear relationship with the security strength of the MDCLCG technique. If n 32 bits, it is polynomial-time unpredictable and secure. As can be seen in Fig.3.2, the hardware design of the MDLCG approach is based on LCG, with the three-operand modulo2n adder serving as the main computational arithmetic block. Four registers, multiplexers, a Square Root Carry Select Adder, and two magnitude comparators make up the MDCLCG design is represented in fig 3.2. The performance of the MDCLCG architecture is influenced by the amount of space utilised in the HCA together with an increase in bit size. As a result, the suggested Square Root Carry Select Adder architectures are used in this part to replace the HCA adder and measure the performance parameters of the MDCLCG. The architecture of the proposed adder is further revised to take into account the MDCLCG method's three-operand modulo-2n adding operation.

The suggested design is synthesised using a 32nm CMOS technology library that is commercially accessible. Moreover, it dissipates less power and has a smaller surface. Moreover, compared to the existing three-operand adder techniques, the proposed adder achieves less area.



4.1: RTL SCHEMATIC:- The RTL schematic is abbreviated as the register transfer level it denotes the blue print of the architecture and is used to verify the designed architecture to the ideal architecture that we are in need of development. The hdl language is used to convert the description or summery of the architecture to the working summery by use of the coding language i.e verilog ,vhdl. The RTL schematic even specifies the internal connection blocks for better analyzing .The figure represented below shows the RTL schematic diagram of the designed architecture.



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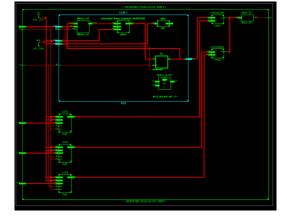


Fig 4.1.1: RTL Schematic of Existed MDCLCG

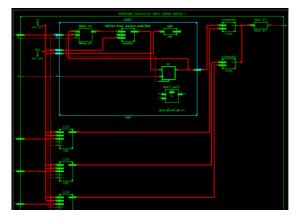


Fig 4.1.2: RTL Schematic of Existed MDCLCG

4.2: TECHNOLOGY SCHEMATIC:- The technology schematic makes the representation of the architecture in the LUT format ,where the LUT is consider as the parameter of area that is used in VLSI to estimate the architecture design .the LUT is consider as an square unit the memory allocation of the code is represented in there LUT s in FPGA.

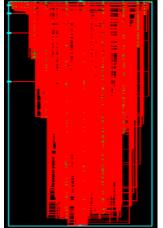


Fig 4.2.1: View Technology Schematic of existed MDCLCG

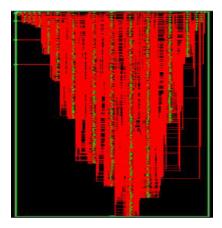


Fig 4.1.1: View Technology Schematic of proposed MDCLCG

4.3 SIMULATION:

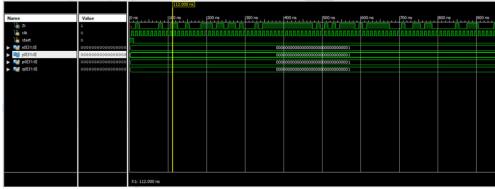


Fig 4.3.1: Simulated Waveforms of existed MDCLCG



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The simulation is the process which is termed as the final verification in respect to its working where as the schematic is the verification of the connections and blocks. The simulation window is launched as shifting from implantation to the simulation on the home screen of the tool ,and the simulation window confines the output in the form of the wave forms. Here it has the flexibility of providing the different radix number systems

		1						_			071.4	67 ns
											071.0	0716
Name	Value	1	100 ns	200 ns	300 ns	400 ns	500 ns	600 ns	700 ns	800 ns		900 ns 1,
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1 clk	0		nnnnnnnn	nnnnnnnn	nnnnnnnn		nnnnnnnn	Innnnnnnn	Innnnnnnn	Innnnn		Innnnnnnn
1 start	0											
▶ 📷 x0[31:0]	000000000000000000000000000000000000000					000000	000000000000000000000000000000000000000	00000001				
▶ 📑 y0[31:0]	000000000000000000000000000000000000000					000000	000000000000000000000000000000000000000	00000001				
▶ 📷 p0[31:0]	000000000000000000000000000000000000000					000000	000000000000000000000000000000000000000	00000001				
▶ 📑 q0[31:0]	000000000000000000000000000000000000000					000000	000000000000000000000000000000000000000	00000001				
		Г										
		1										
		X1	X1: 871.667 ns									

Fig 4.3.2: Simulated Waveforms of proposed MDCLCG

The simulation is the process which is termed as the final verification in respect to its working where as the schematic is the verification of the connections and blocks. The simulation window is launched as shifting from implantation to the simulation on the home screen of the tool and the simulation window confines the output in the form of the wave forms. Here it has the flexibility of providing the different radix number systems.

Parameter comparison

	Parameter	Existed MDCLCG	Proposed MDCLCG					
	No of LUTs	715	646					
Table 4.3.3: LUT's comparison								

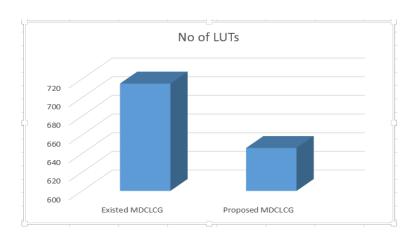


Fig 4.4.4: LUT comparison Bar graph

V.CONCLUSION & FUTURE SCOPE

Modified Dual-CLCG method involves dual coupling of four LCGs that makes it more secure than LCG based PRBGs. However, It is reported that this method has the drawback of generating pseudorandom bit at large area and more delay. Proposed architecture of the new modified dual- CLCG method using square root carry select adder is significantly



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reduced the area of the design. The proposed architecture of the modified dual- CLCG method is prototyped on the commercially available FPGA devices and the results are captured in real-time using Xilinx chip scope for validation. Based on the performance analysis in terms of hardware complexity, randomness and security, it is observed that 32- bit hardware architecture of the proposed modified dual-CLCG method is optimum and can be useful in the less area of hardware security and IoT applications, cryptography and PRBG applications.

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