



# **Enhanced Beam forming Designs for Multiuser MIMO-OFDM Interference Channels**

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**ABSTRACT:** Three enhanced beamforming designs for multiuser multiple input and multiple output with orthogonal frequency-division multiplexing, where the transmit and receive beamformers are obtained iteratively with closed-form steps. In the first case, the transmit (Tx) beamformers are set and the receive (Rx) beamformers are calculated. It works by projecting the Tx beamformers into a null space of appropriate channels. This eliminates one interference term for each user. Then the Rx beamformer for each user maximizes its instantaneous signal-to-noise ratio (SNR) while satisfying an orthogonality condition to eliminate the remaining interferences. The second case is jointly optimizing the Tx and Rx beam formers from constrained SNR maximization. It uses the results from the first case. The third case is also for joint optimization of Tx–Rx beamformers but combines constrained SNR and signal-to-interference-plus-noise ratio maximization. All cases can include a linear constellation precoder for extracting multipath diversity. The space frequency block codes are used for MIMO construction and SVD technique is used for reducing the complexities during beamforming. Also the system includes a successive interference canceller at the receiver side to reduce more interference. The system is compared with the existing systems based on BER, sumrate, SER and transmission rate. Our simulations demonstrate faster beamforming, improved error performance, higher transmission rate and has the ability to extract multipath diversity. Which makes the system effective than other existing systems.

**KEYWORDS:** Beamforming; multiple input multiple output antennas ; SINR;SNR;SVD

## **I. INTRODUCTION**

Beamforming for a multiuser MIMO interference channel is for communications between pairs of terminals where there are several pairs sharing the spectrum simultaneously. Each multi-antenna transmitter strives to direct its data to only one multielement receiver in the presence of interference from all the other users' transmitters. This is a recent subject, and it is different from multiuser uplink and downlink beamforming. The use of an SINR constraint, where independent data streams are transmitted from a multi-antenna base exploited uplink-downlink SINR duality. But none of these can be applied to the more complicated interference channel which is the subject of this paper. In this paper, the signal processing operations are referring to beamforming as a vector operator. Therefore, only singlestream data transmission is considered. So multi-stream data transmission for each user, usually discussed in IA design, is not used here. all users have multiple transmit and receive antennas. The joint transmit beamformer (Tx-BF) and receive beamformer (Rx-BF) design for minimum SINR maximization in a MIMO interference channel system has been proven to be a strongly NP-hard problem. It is clear that there are several different approaches to the beamforming problem. The ultimate metric is a practicable digital communications performance, but it is not yet possible to optimize this directly. Instead, the optimization of some analogue channel performance functions is followed by a calculation of some aspect of the associated communications performance usually an information theoretic capacity or throughput rate, along with some error performance.

Along these lines, this paper presents three new beamforming design cases. First, a constrained SNR maximization is sought in which the Tx-BFs for all the users are acquired by deploying the null-space of an appropriate channels matrix. This null-space assignment for Tx-BFs eliminates one term of interference at each receiver. The remaining interference terms at each receiver can be eliminated by means of orthogonal vectors. The Rx-BF of each user, in this case, has a closedform solution if its norm is one. The second case is joint Tx–Rx beamformer design for constrained SNR



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maximization where only the Rx-BF at each receiver terminal nulls out all the interference. This problem leads to a multi-objective optimization which can be solved iteratively because it has guaranteed convergence. The third case is joint constrained SNR and SINR maximization. This problem has a solution because its corresponding vector field is nonexpansive. It can be considered as a new formulation for LI-SP-SINR. Also the system includes a successive interference canceller at the receiver side to reduce the further interference. Space frequency based block codes are used at the transmitting side which increasing the transmission rate and diversity.

## II. RELATED WORK

In [2] they proposed about the Cooperative algorithms for MIMO interference channels. This paper proposes three generalization of IA (interference alignment) for the multiple antenna interference channel with multiple users that account for colored noise, which models uncoordinated interference. First a minimum interference plus noise leakage algorithm is presented and shown to be equivalent to previous subspace methods when noise is spatially white or negligible. This algorithm results in orthonormal precoders that are desirable for practical implementation with limited feedback. A joint minimum mean square design that jointly optimizes the transmit precoders and receive spatial filters is then processed. Finally, a maximum signal to interference plus noise ratio (SINR) algorithm is developed and proven to converge unlike previous maximum SINR algorithms. In [3] Authors proposed about The feasibility of interference alignment over measured MIMO-OFDM channels. In this paper, interference alignment (IA) has been shown to achieve the maximum achievable degrees of freedom in the interference channel. This results in sum rate scaling linearly with the number of users in the high signal to noise ratio regime. Linear scaling is achieved by precoding the transmitted signals to align interference subspaces at the receivers given channel knowledge of all transmit receive pairs, effectively reducing the number of discernible interferences. The theory of IA was derived under assumptions about the richness of scattering in the propagation channel; practical channels do not guarantee such ideal characteristics. This paper presents the first experimental study of IA in measured multiple-input-multiple output orthogonal frequency-division-multiplexing (MIMO-OFDM) interference channels. In [4] authors proposed MSE-based transceiver designs for the MIMO interference channel. Interference alignment (IA) has evolved as a powerful technique in the information theoretic framework for achieving the optimal degrees of freedom of interference channel. In practical systems, the design of specific interference alignment schemes is subject to various criteria and constraints. In this paper novel transceiver schemes for the MIMO interference channel based on the mean square error (MSE) criterion is considered. The objective is to optimize the system performance under a given feasible degree of freedom. Both total MSE and the maximum per user MSE are chosen to be objective functions to minimize. It shows that the joint design of transmit precoding matrices and receiving filter matrices with both objectives can be realized through efficient iterative algorithms. The proposed scheme outperforms the existing IA schemes in terms of BER performance. In [5] they used weighted sum rate maximization by which the constrained problem was converted to an unconstrained problem and solved by gradient descent algorithm. It requires more feedback compared to our proposed system. Complexity is high and only single stream data transmission is considered because multi stream data transmission for each user in IA design is not used here.

## III. EXISTING SYSTEM

Presents three beamforming design cases. First, a constrained SNR maximization is sought in which the Tx-BFs for all the users are acquired by deploying the null-space of an appropriate channels matrix. This null-space assignment for Tx-BFs eliminates one term of interference at each receiver. The remaining interference terms at each receiver can be eliminated by means of orthogonal vectors. The Rx-BF of each user, in this case, has a closed form solution if its norm is one. The second case is joint Tx-Rx beamformer design for constrained SNR maximization where only the Rx-BF at each receiver terminal nulls out all the interference. This problem leads to a multi-objective optimization which can be solved iteratively because it has guaranteed convergence. The third case is joint constrained SNR and SINR maximization. This problem has a solution because its corresponding vector field is nonexpansive. It can be considered as a new formulation for LI-SP-SINR. For these three cases, the minimum required number of antennas is derived as a part of the formulation. Finally, the required feedback rates are computed for each approach and compared with those of existing beamforming schemes. Two out of three proposed cases require less feedback if the same number of antennas is considered. These methods are also computationally simple, but at the price of one more antenna element at each terminal.

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In order to further improve the digital communications error performance (without compromising the sum rate performance), the system formulation can include a precoder matrix before the transmit beamformer and a sphere decoder (SD) following the receiver beamformer. This is to allow multipath diversity gain from the OFDM system. This precoder is a fixed matrix that does not need instantaneous channel knowledge, although it does need knowledge of some channel characteristics for optimal Deployment. It has high SER less transmission rate.

### A. System model:

The communication situation related model can be considered in the form of system model. There are  $K$  pairs of multi-antenna terminals which are striving to share simultaneously the spectrum in time and space. The channel is modeled as 1) a tapped delay line ( $L + 1$  taps) according to the IEEE 802.11n propagation model or 2) a single-tap flat fading channel with a perfect spatial correlation matrix. The first channel model justifies both the MIMO-OFDM configuration and deploying the multipath precoder in this paper. The second channel model is used only to compare the performance of the system. The  $K$  users all have  $N_t$  transmit antennas and  $N_r$  receive antennas, and all users utilize each of the  $P$  subchannels. The transmit beamformers for the  $i$ th user at the  $p$ th subcarrier are written  $V_i(P)$ , the receive beamformers are  $U_i(P)$  the MUX block stacks  $P$  samples of  $S_i$ . where  $S_i$  is the input symbol stream of user  $i$ . The users data symbols are assumed to be mutually independent. we consider an optimal matrix  $\varphi$  to extract multi-path diversity

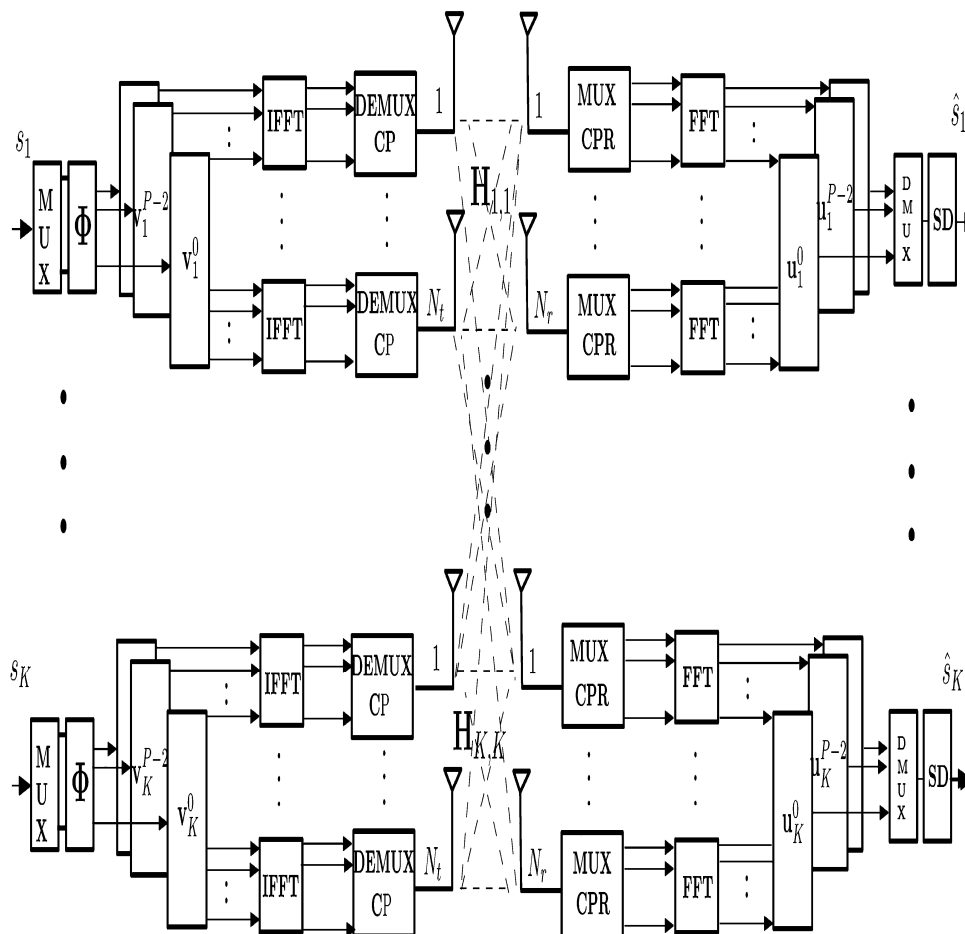


Fig.1. Existing system

### B. Optimal RX-BFS for constrained SNR maximization when the TX-BFS are known:

The Tx-BFs are found from the null space of an appropriate set of channels, and then the optimal Rx-BFs are sought. For  $K \in \{2(n+1):n \in \mathbb{N}\}$ , where  $N$  denotes positive integers, the beamformer  $V_i$  is obtained by eq. (1).

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$$V_i = N(H_{k+1-i}, i) \text{ where } N(A) \triangleq \{x | Ax = 0, \|x\| = 1\} \quad \text{eq. (1)}$$

Is an orthonormal basis for the null space of A. For  $K \in \{2n+1: n \in N\}$ , The next step is to determine  $U_i$  such that it maximizes the signal-to-noise ratio (SNR) of the  $i$ th user (i.e., after the Rx-BF) while suppressing the  $K-2$  remaining interference terms. This optimization problem is denoted P for the first receiver as an example,

$$P: \max_{u_1 \in C^{N_r}} \frac{U_1^H H_{1,1} V_1 V_1^H H_{1,1}^H U_1}{U_1^H U_1} \quad \text{eq. (2)}$$

So P is a constrained SNR maximization formulation where maximization over a quasi-convex object function with affine constraints is sought. The maximizing of the ratio of quadratic forms is a known problem with an eigen solution. But here the difference is that the P has constraints. This different problem leads to a different solution. These constraints force the interference, for the first user as an example, to be eliminated. To solve P, its Lagrangian function is needed

### C. Joint RX-BF and TX-BF for constrained SNR maximization:

Joint Tx-BF and Rx-BF are designed for the constrained SNR maximization problem by using the extended alternating optimization (EAO) algorithm for a multi-objective optimization. Consider the following optimization problem,

$$\min_{x \in \Omega_j} \tau(x) = \tau_1(x_1, \dots, x_k) + \tau_k(x_1, \dots, x_k) \quad \text{eq. (3)}$$

Where  $x = [x_1, \dots, x_k]^T$  and  $\Omega_j$  is a feasible set. Generally, solving such a nonlinear constrained optimization problem is difficult. However, if firstly for each objective function,  $\tau_i, i = 1, \dots, k$  there is a unique global minimizer with  $x_i$  respect to for fixed  $x_1, \dots, x_{i-1}, x_{i+1}, x_k$ , then EAO approximates the difficult problem's solution by simultaneous solving of the following K problems. optimal solution can be represented as  $x_i = l_i(x_1, \dots, x_{i-1}, x_{i+1}, x_k)$ . We show that the EAO algorithm can be deployed for joint V and U design for constrained SNR problem. Assume that the  $V_2, \dots, V_k$  are assigned arbitrarily at the first iteration, then the joint constrained SNR problem for the first user is formulated as,

$$G: \max_{v_1 \in C^{N_t}, u_1 \in C^{N_r}} U_1^H H_{1,1} V_1 V_1^H H_{1,1}^H U_1 \quad \text{eq. (4)}$$

### D. TX-BF and RX-BF design for joint constrained SNR maximization and SINR maximization :

Each Rx-BF nulls its interference and then this solution is inserted to the constrained SNR objective function which yields the Tx beamformer. In this section, multiobjective optimization by the fixed point method is applied. Instead of G, which is optimization w.r.t.  $V_1$  and  $U_1$ , define the problem  $G_1$  as,

$$G: \max_{v_1 \in C^{N_t}} U_1^H H_{1,1} V_1 V_1^H H_{1,1}^H U_1 \quad \text{eq. (5)}$$

This problem is maximization w.r.t.  $V_1$  only.

## IV. PROPOSED SYSTEM

In the proposed system we enhancing the Beamforming designs for multi user MIMO-OFDM interference channel with successive interference canceller at the receiver side and this system model is considering for the three cases of the beamforming designs. Also space frequency block codes are using for the signal transmission which can avoid the limitation caused by the time domain analysis. The system implements singular value based decomposition technique in analysis.

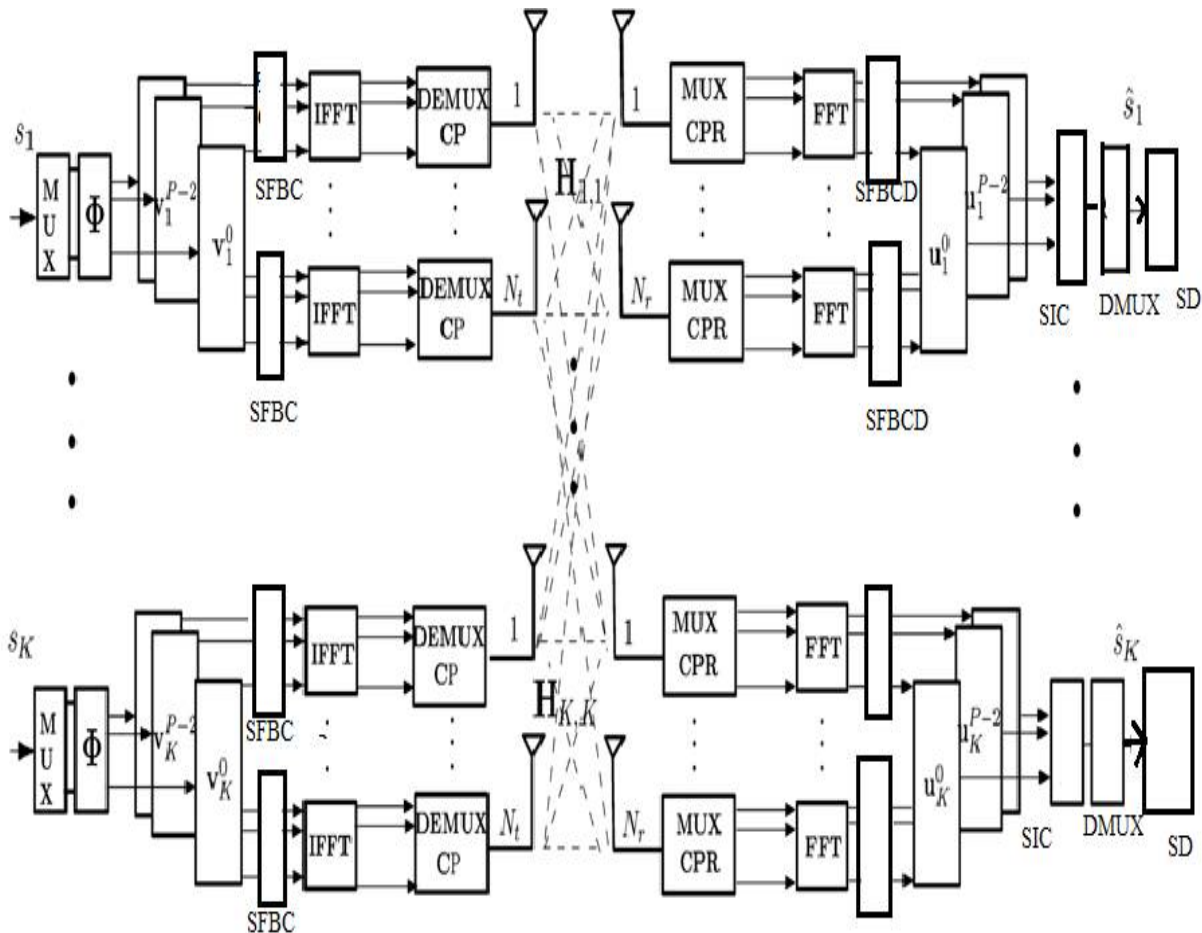


Fig.2. Proposed system

It is proposed to use defected ground structures for improving the isolation as compared to electromagnetic band gap structure. Monopole antenna structure is varied to reduce the self-interference and to analyse the effect in radiation pattern. In order to increase the bandwidth feed to feed transition technique is used. A microstrip-slot transition is a structure that uses a microstrip line on one side of a planar dielectric substrate and a slot line in the ground so that a signal is passed between the two sides. In order to perform this function with minimal power losses, the microstrip line and the slotline have to be orthogonal to each other and use suitable reactive terminations at their ends. The system SER is low and transmission rate is high compared to the existing system.

## V. SIMULATION RESULTS

For simplicity, all the users use QPSK in the evaluation of BER performance. In terms of performance, symbol-wise and frame-wise detection at the receiver have the same result. Fig. 3 shows the BER for  $K=3$  users over IEEE 802.11n channel model B. The joint multi-objective Tx-BF and Rx-BF design (approach 3) has the best performance, then the joint single-objective Tx-BF and Rx-BF design (approach 2), and finally the individual design (approach 1). This simulation also demonstrates that for this MIMO channel for WLAN with 9 paths, using the LCP matrix still improves the system performance. Finally it is recalled that the complexity of SD decoding at the receiver increases with the constellation size but not with the number of antennas. The key benefit of the presented beamforming approaches is their computational simplicity.

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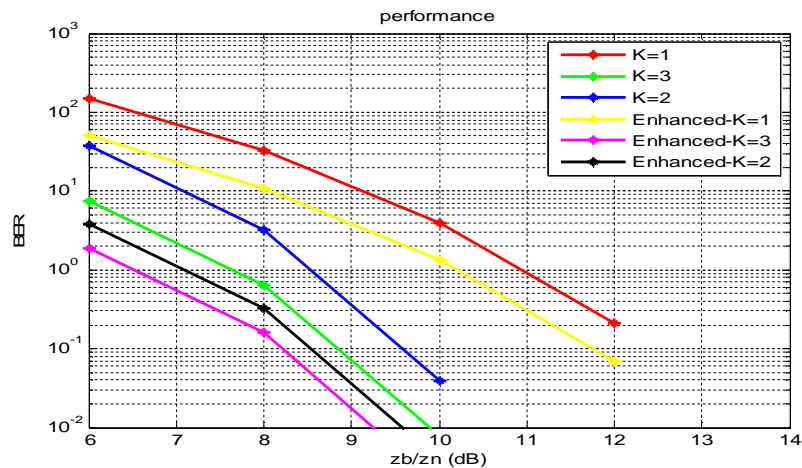


Fig. 3. The BER performance of approaches 1, 2, 3 for K=3 with (w) or without (w.o) deploying LCP precoder and sphere decoder with IEEE 802.11n channel model B.

In terms of performance symbol-wise and frame-wise detection at the receiver have the same result. Fig. 3 shows the BER for K=3 users over IEEE 802.11n channel model B. The complexity of the SD decoding at the receiver increases with the constellation size but not with the number of antennas.

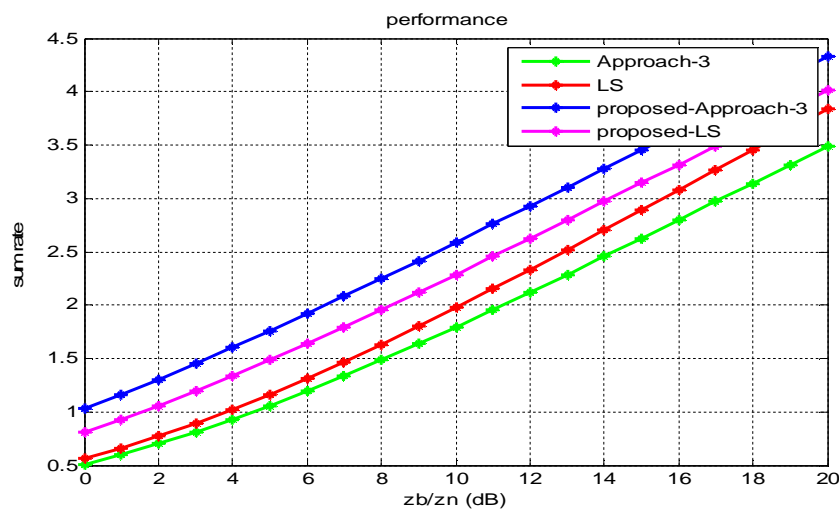


Fig. 4. The sum rate performance of the approach 3 and LS transmit beamformer design for K=4 users with IEEE 802.11n channel model A.

Fig. 4 illustrates that it also has better sum-rate performance and also lower BER performance. The sum rate in bits/s/Hz is computed by  $\sum_{i=1}^K \log_2(1 + \text{SINR}_i)$ . Here, the channel is the IEEE 802.11n channel model A, which is a flat-fading MIMO channel with average path gain of 0 dB, angular spreads  $\text{AS} = 40^\circ$  at the transmitter and receiver, mean angle of departure is  $45^\circ$  and mean angle of arrival is  $45^\circ$ .

The BER performance and computational complexity of proposed methods have been compared with existing known methods, such as ZF, LI and sum-rate maximization. From Fig. 5, approaches 2 and 3 have better performance than ZF with selection and also much better than LI. The channel for this simulation is flat fading in time domain denoted by  $G(0, 1)$ . Fig. 6 compares the sum rate maximization by EAO with sum rate maximization by gradient descent method. From Fig. 5, the performance loss is only 0.8 dB. The computational complexity of this gradient method is  $O(NK^2M^3)$  where  $M = \max(N_t, N_r)$ . However, the computational complexity of all closed-form methods

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presented here is  $O(NKM^3)$ . The required CPU times (not shown) indicate that our proposed method is five times faster for  $M=4$  and  $K=3$ .

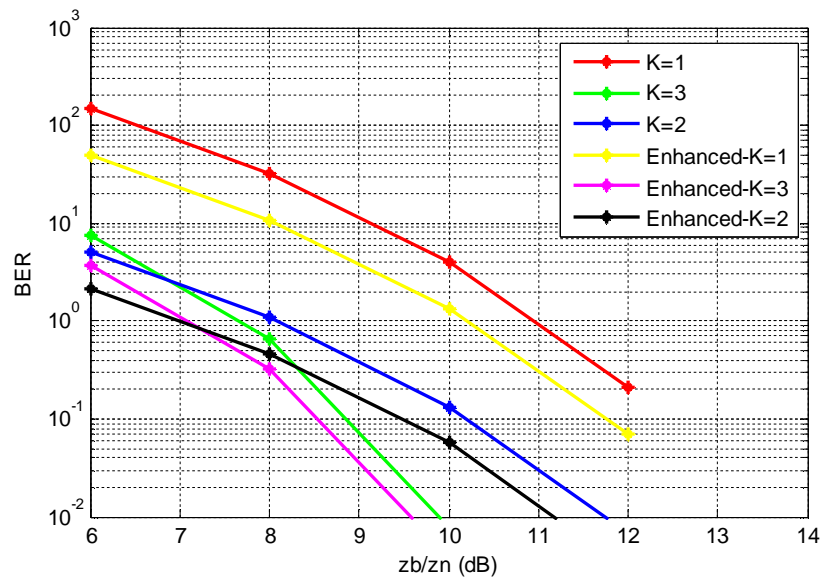


Fig. 5. The BER performance of ZF with selection, LI and proposed approaches 2 and 3 for  $K=3$  and  $G(0, 1)$  channels.

Robustness to imperfect channel information can be readily gauged in the usual way by modeling the channel with  $H_{i,j} = \rho H_{i,j} + \sqrt{(1 - \rho^2)}W$ , where  $W$  is a zero mean, unit variance complex Gaussian random matrix

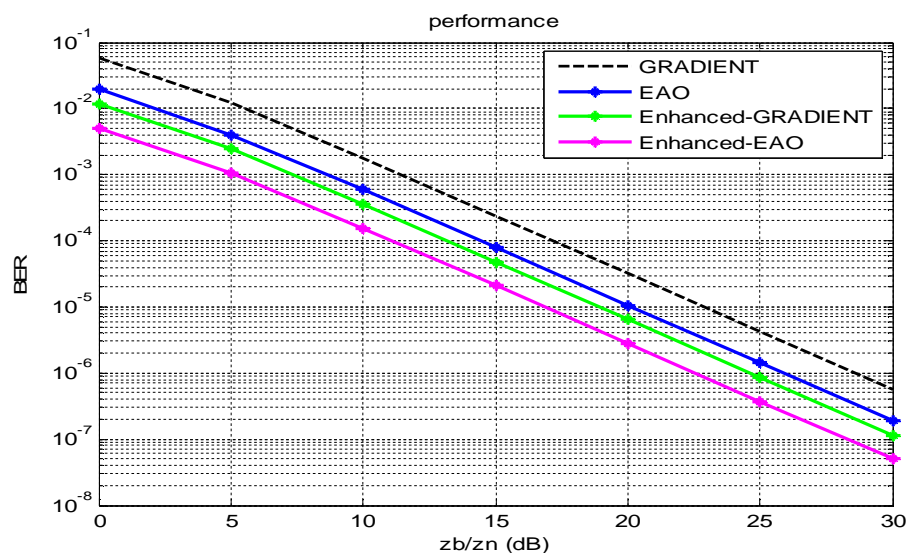


Fig. 6. The BER performance comparison of gradient method and proposed EAO, with sum rate objective function, for  $K=3$  and  $G(0, 1)$  channels

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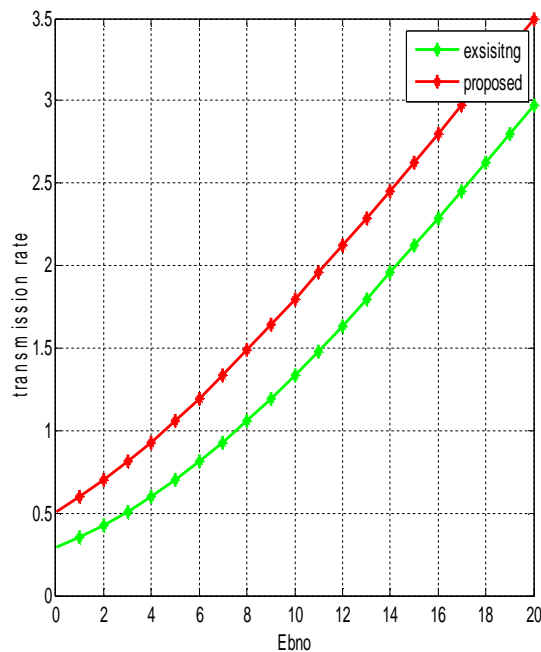


Fig. 7. Transmission rate comparison

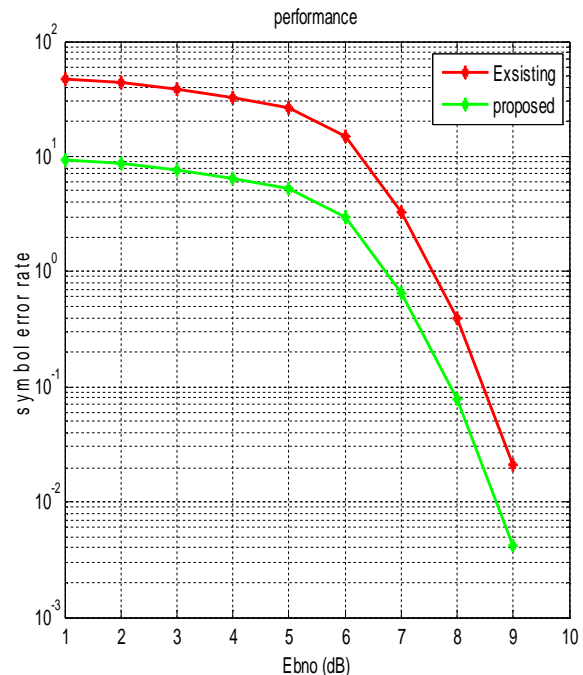


Fig. 8. SER comparison

## VI. CONCLUSION

We can enhance the performance of a multi user MIMO-OFDM interference channel by the use of space frequency block code at the transmitting side and a successive noise canceller at the receiver side. Singular value based decomposition can be used for reducing the complexity in beamforming. Since the system is converting from time domain to frequency domain complexity can be reduced. We can achieve high transmission rate, high diversity and error free fast transmission than the existing system.

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